

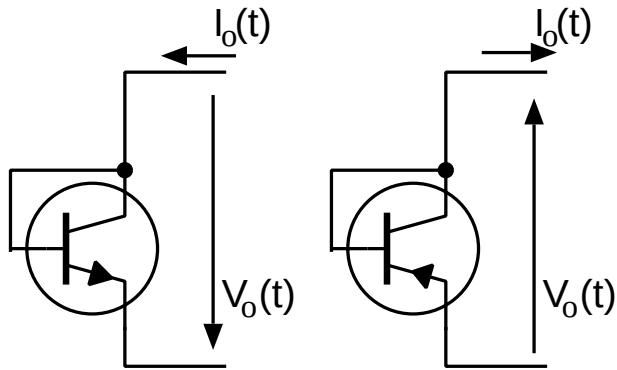
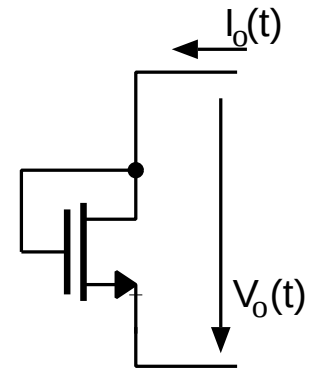
Tema 2. Bloques Constructivos

- Conexion Diodo
- Fuente de Corriente
- Espejo de Corriente
- Repaso transistor bipolar
- Repaso transistor MOSFET

Bibliografia

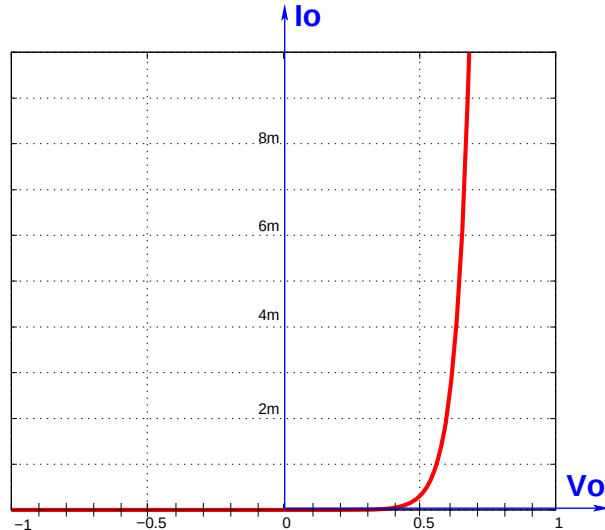
“VLSI Design Techniques for Analog and Digital Circuits”, Randall Geiger, Noel Strader, N. R. Strader, McGraw-Hill

Conexión diodo (*Diode Connected Transistor*)



Modelo DC (Estático)
Supongamos el bjt en activa directa:

$$I_o = I_s e^{V_o/V_t}$$



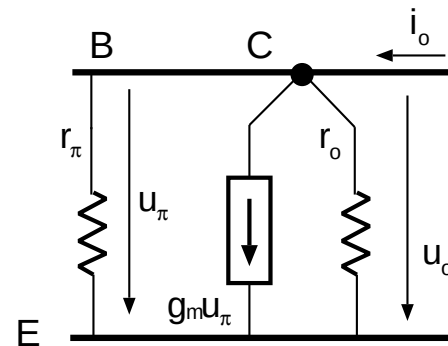
Modelo AC(dinámico)

$$u_o(t) = R i_o(t)$$

metodo A

$$R = \frac{\partial V_o}{\partial I_o} = \frac{V_t}{I_s e^{V_o/V_t}} = \frac{V_t}{I_o}$$

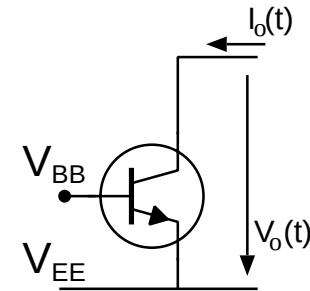
metodo B



$$R = \frac{1}{g_m} || r_{\pi} || r_o$$

Fuente de Intensidad (*Current Source*)

Fuente de corriente simple con bipolar
(*Bipolar Simple Current Source*)

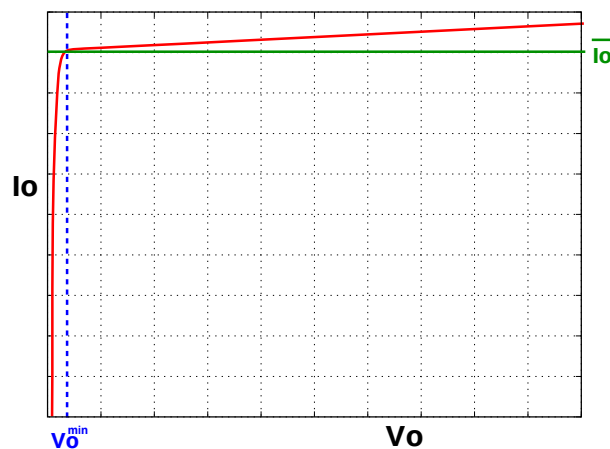


Modelo DC (Estático)

Supongamos el bjt en activa directa:

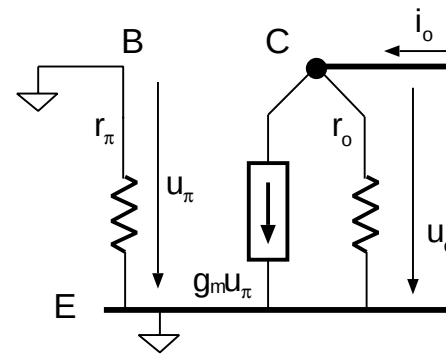
$$I_o = I_s e^{V_{BE}/V_t}$$

$$V_{BE} = V_{BB} - V_{EE} = cte$$



Modelo AC(dinámico)

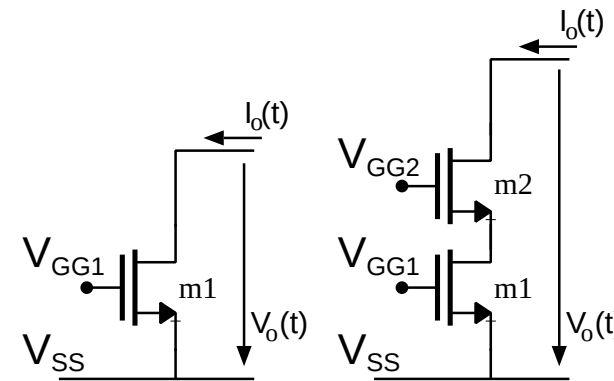
$$u_o(t) = R i_o(t)$$



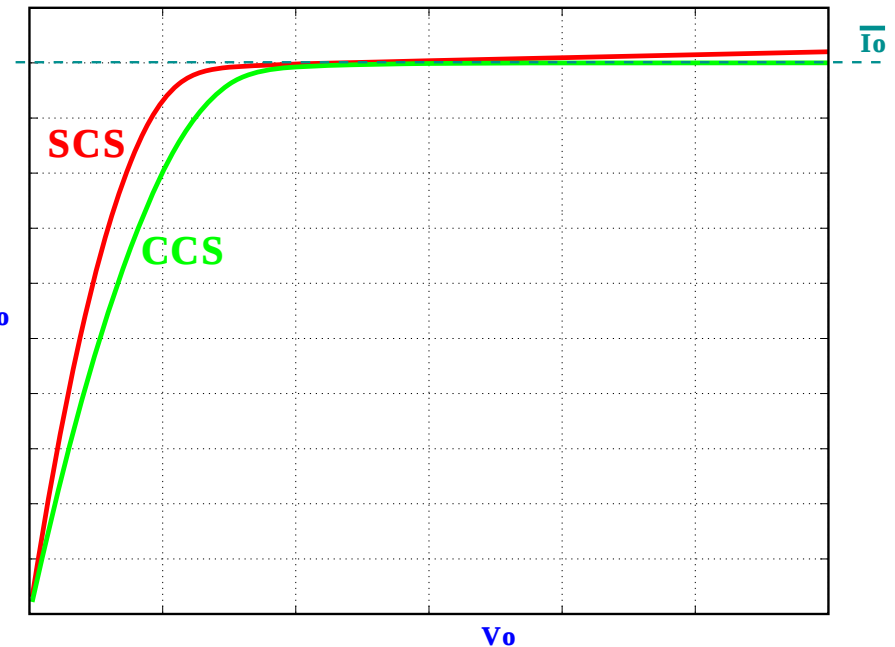
$$R = r_o$$

Fuente de corriente con MOSFET (MOSFET Current Source)

Simple Current Source



Cascode Current Source



Modelo DC (Estático)
Supongamos el MOSFET en saturación:

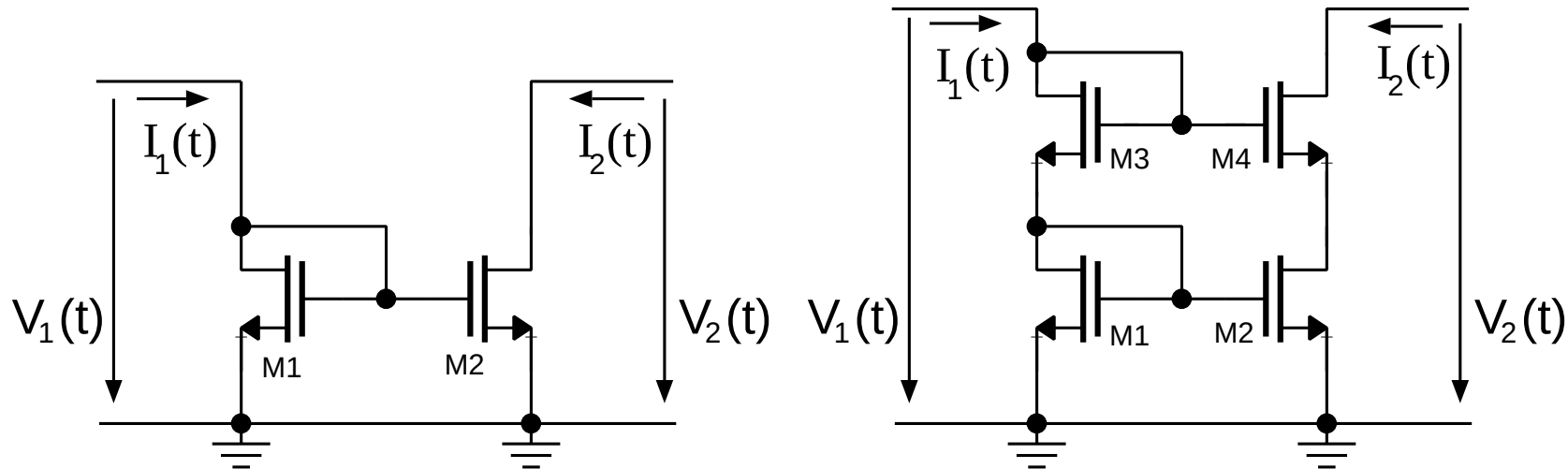
$$I_{d1} = \frac{1}{2} K \frac{W}{L} (V_{gs1} - V_{th1})^2$$

$$V_{gs1} = V_{GG1} - V_{SS}$$

Modelo AC(dinámico)
 $u_o(t) = R i_o(t)$

SCS: $R = r_{o1}$
CCS: $R \simeq r_{o1} g_{m2} r_{o2}$

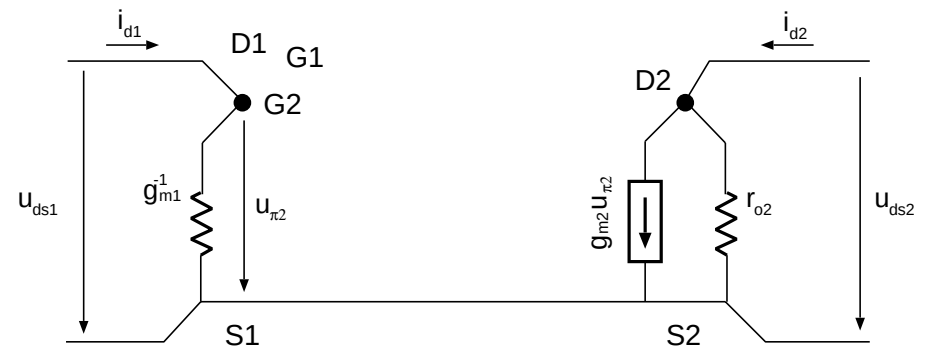
Espejo de Corriente (*Current Mirror*)



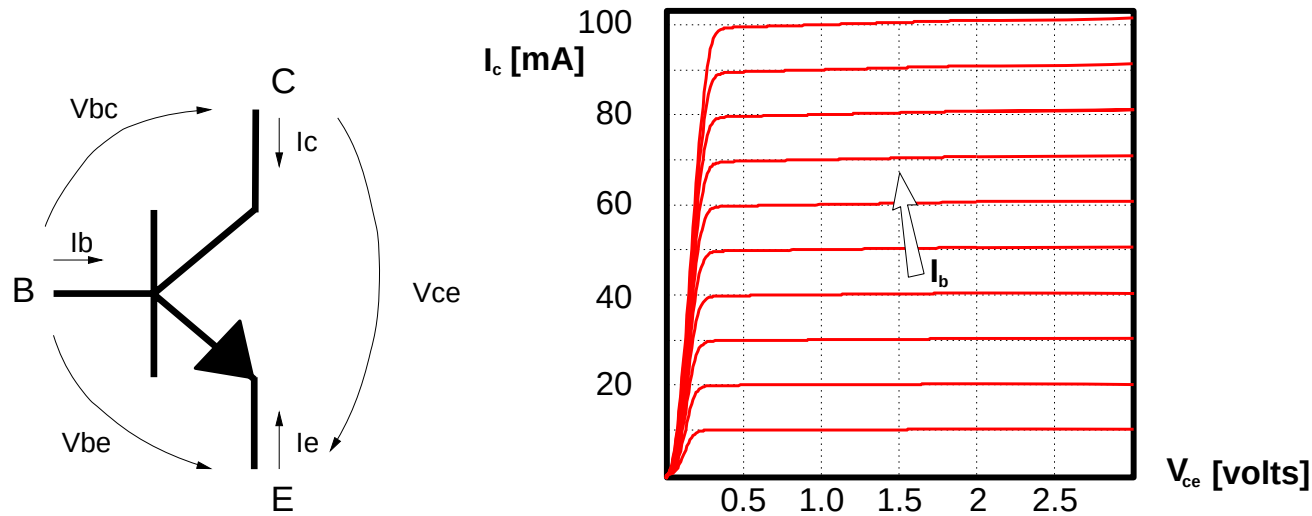
Modelo DC (Estático)
Supongamos MOSFETs en saturación:

$$\frac{\overline{I_1}}{\overline{I_2}} = \frac{(W/L)_1}{(W/L)_2}$$

Modelo AC(dinámico)



Modelo Transistor Bipolar



Modelo DC

Ecuaciones de Ebers-Moll

$$I_c = +I_s \left[\exp \left(\frac{V_{be}}{V_t} \right) - 1 \right] - \frac{I_s}{\alpha_r} \left[\exp \left(\frac{V_{bc}}{V_t} \right) - 1 \right]$$

$$I_e = -\frac{I_s}{\alpha_f} \left[\exp \left(\frac{V_{be}}{V_t} \right) - 1 \right] + I_s \left[\exp \left(\frac{V_{bc}}{V_t} \right) - 1 \right]$$

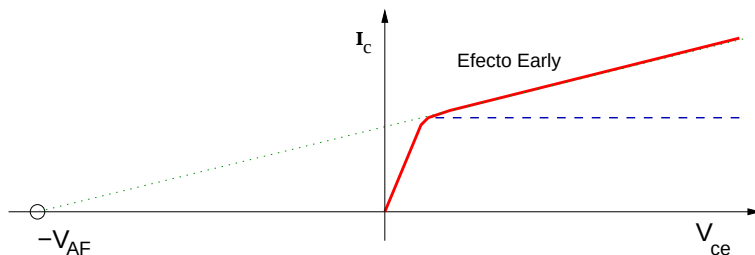
Suponiendo **activa directa**: $V_{be} \gg$ y $V_{bc} \ll$
y definiendo la ganancia directa en corriente
 $\beta_f = \alpha_f / (1 - \alpha_f)$:

$$I_c = I_s \exp\left(\frac{V_{be}}{V_t}\right)$$

$$I_c = \beta_f I_b$$

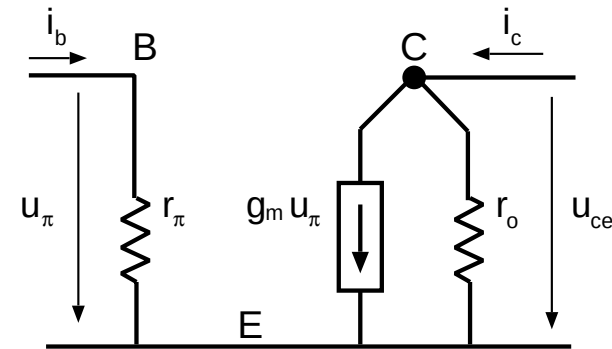
Efecto Early

$$I_c \simeq I_s \exp\left(\frac{V_{be}}{V_t}\right) \left(1 + \frac{V_{ce}}{V_{AF}}\right)$$



Modelo AC

Suponiendo el transistor polarizado en activa directa



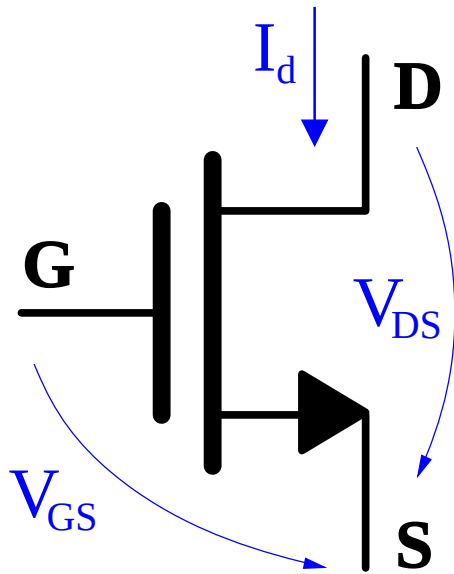
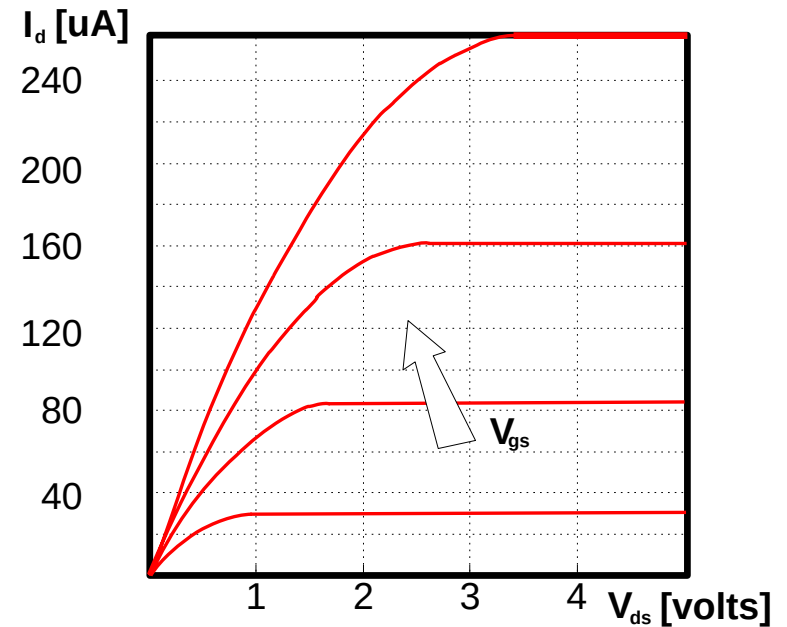
La dependencia de los parametros AC con el nivel DC es:

$$r_o = \frac{V_{AF}}{\bar{I}_c}$$

$$\frac{1}{g_m} = \frac{V_t}{\bar{I}_c}$$

$$r_\pi = \frac{h_{fe} V_t}{\bar{I}_c}$$

Modelo Transistor MOSFET



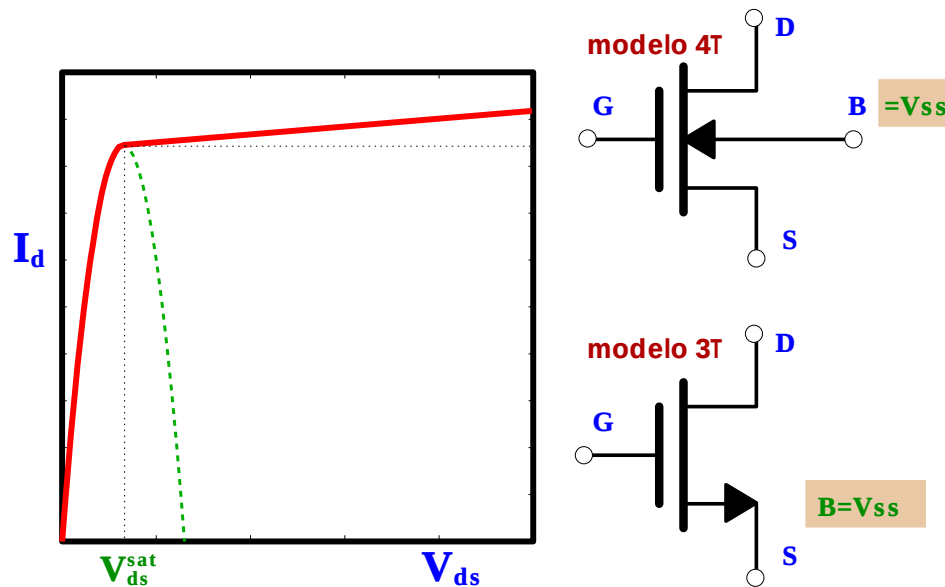
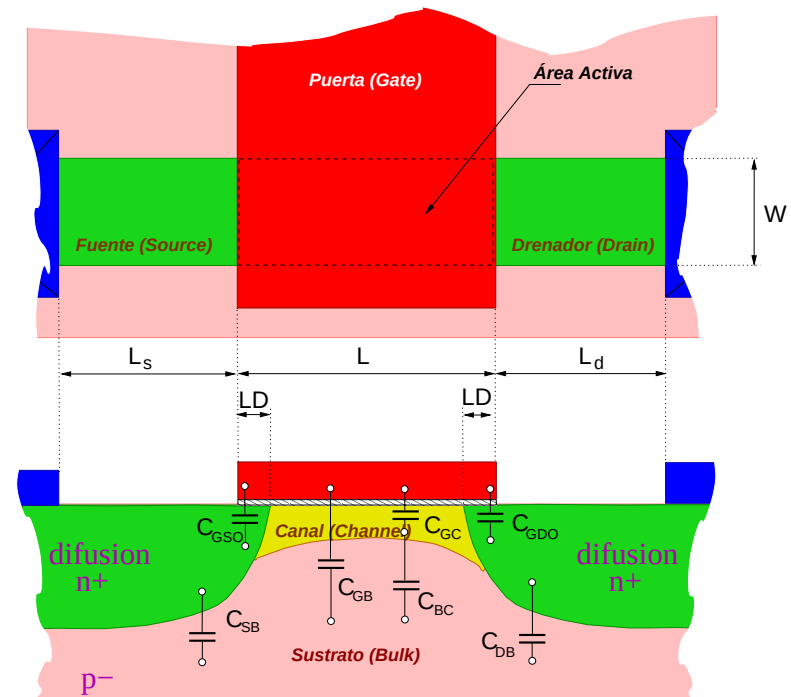
Modelo DC

En la **zona de saturación** tenemos que:

$$I_d = \frac{1}{2} K \left[\frac{W}{L} \right] [V_{gs} - V_{th}]^2$$

siempre que $V_{gs} > V_{th}$ y $V_{ds} > V_{gs} - V_{th}$

- $K \left[\frac{\mu A}{V^2} \right]$ **Parámetro de transconductancia** (*transconductance parameter*). $K = \mu_o C_{ox}$
- $\left[\frac{W}{L} \right]$ **relación de aspecto** (*aspect ratio*)
- V_{th} **tensión umbral** (*threshold voltage*)



Modulación del Canal

$$I_d = \frac{1}{2} K \left[\frac{W}{L} \right] \left[V_{gs} - V_{th} \right]^2 (1 + \lambda V_{ds})$$

Efecto Sustrato

$$V_{th} = V_{to} + \gamma \left[\sqrt{\Phi_f + V_{sb}} - \sqrt{\Phi_f} \right]$$

Modelo AC

$$g_m = \sqrt{2 K \left[\frac{W}{L} \right] I_d}$$

$$\mathbf{r}_o = 1/(\lambda \mathbf{I}_d)$$

$$g_{mb} = \eta \, g_m$$

